

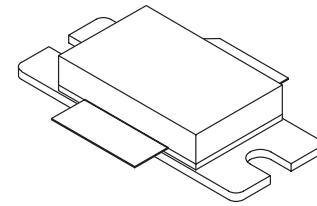
The RF MOSFET Line
RF Power Field Effect Transistors
N-Channel Enhancement-Mode Lateral MOSFETs

MRF18090B
MRF18090BS

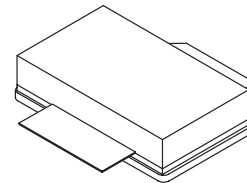
Designed for GSM and EDGE base station applications from frequencies up to 1.9 to 2.0 GHz. Suitable for FM, TDMA, CDMA and multicarrier amplifier applications. To be used in class AB for GSM and EDGE cellular radio applications.

- GSM and EDGE Performances, Full Frequency Band
Power Gain — 13.5 dB (Typ) @ 90 Watts (CW)
Efficiency — 45% (Typ) @ 90 Watts (CW)
- Internally Matched, Controlled Q, for Ease of Use
- High Gain, High Efficiency and High Linearity
- Integrated ESD Protection: Class 2 Human Body Model, Class M3 Machine Model
- Ease of Design for Gain and Insertion Phase Flatness
- Capable of Handling 10:1 VSWR, @ 26 Vdc, 90 Watts (CW) Output Power
- Excellent Thermal Stability
- Characterized with Series Equivalent Large-Signal Impedance Parameters

90 W, 1.90 – 1.99 GHz, 26 V
LATERAL N-CHANNEL
BROADBAND
RF POWER MOSFETS



CASE 465B-02, STYLE 1



CASE 465C-01, STYLE 1

MAXIMUM RATINGS

| Rating | Symbol | Value | Unit |
|--|-----------|-------------|------------------------------------|
| Drain-Source Voltage | V_{DSS} | 65 | Vdc |
| Gate-Source Voltage | V_{GS} | +15, -0.5 | Vdc |
| Total Device Dissipation @ $T_C = 25^\circ\text{C}$ Derate above 25°C | P_D | 250 1.43 | Watts $\text{W}/^\circ\text{C}$ |
| Storage Temperature Range | T_{stg} | -65 to +200 | $^\circ\text{C}$ |
| Operating Junction Temperature | T_J | 200 | $^\circ\text{C}$ |

THERMAL CHARACTERISTICS

| Characteristic | Symbol | Max | Unit |
|--------------------------------------|-----------------|-----|---------------------------|
| Thermal Resistance, Junction to Case | $R_{\theta JC}$ | 0.7 | $^\circ\text{C}/\text{W}$ |

NOTE – **CAUTION** – MOS devices are susceptible to damage from electrostatic charge. Reasonable precautions in handling and packaging MOS devices should be observed.

ELECTRICAL CHARACTERISTICS ($T_C = 25^\circ\text{C}$ unless otherwise noted)

| Characteristic | Symbol | Min | Typ | Max | Unit |
|--|---------------|---|------|-----|-----------------|
| OFF CHARACTERISTICS | | | | | |
| Drain–Source Breakdown Voltage ($V_{GS} = 0\text{ Vdc}$, $I_D = 100\ \mu\text{Adc}$) | $V_{(BR)DSS}$ | 65 | — | — | Vdc |
| Zero Gate Voltage Drain Current ($V_{DS} = 26\text{ Vdc}$, $V_{GS} = 0$) | I_{DSS} | — | — | 10 | μAdc |
| Gate–Source Leakage Current ($V_{GS} = 5\text{ Vdc}$, $V_{DS} = 0$) | I_{GSS} | — | — | 1 | μAdc |
| ON CHARACTERISTICS | | | | | |
| Gate Quiescent Voltage ($V_{DS} = 26\text{ Vdc}$, $I_D = 750\text{ mAdc}$) | $V_{GS(Q)}$ | 2.5 | 3.7 | 4.5 | Vdc |
| Drain–Source On–Voltage ($V_{GS} = 10\text{ Vdc}$, $I_D = 1\text{ Adc}$) | $V_{DS(on)}$ | — | 0.1 | — | Vdc |
| Forward Transconductance ($V_{DS} = 10\text{ Vdc}$, $I_D = 3\text{ Adc}$) | g_{fs} | — | 7.2 | — | S |
| DYNAMIC CHARACTERISTICS | | | | | |
| Reverse Transfer Capacitance ($V_{DS} = 26\text{ Vdc}$, $V_{GS} = 0$, $f = 1\text{ MHz}$) | C_{rss} | — | 4.2 | — | pF |
| FUNCTIONAL TESTS (In Motorola Test Fixture) | | | | | |
| Common–Source Amplifier Power Gain @ 90 W (1) ($V_{DD} = 26\text{ Vdc}$, $I_{DQ} = 750\text{ mA}$, $f = 1930 - 1990\text{ MHz}$) | G_{ps} | 12 | 13.5 | — | dB |
| Drain Efficiency @ 90 W (1) ($V_{DD} = 26\text{ Vdc}$, $I_{DQ} = 750\text{ mA}$, $f = 1930 - 1990\text{ MHz}$) | η | 40 | 45 | — | % |
| Input Return Loss (1) ($V_{DD} = 26\text{ Vdc}$, $P_{out} = 90\text{ W CW}$, $I_{DQ} = 750\text{ mA}$, $f = 1930 - 1990\text{ MHz}$) | IRL | 10 | — | — | dB |
| Output Mismatch Stress ($V_{DD} = 26\text{ Vdc}$, $P_{out} = 90\text{ W CW}$, $I_{DQ} = 750\text{ mA}$ VSWR = 10:1, All Phase Angles at Frequency of Tests) | Ψ | No Degradation In Output Power Before and After Test | | | |

(1) To meet application requirements, Motorola test fixtures have been designed to cover the full GSM1900 band, ensuring batch–to–batch consistency.

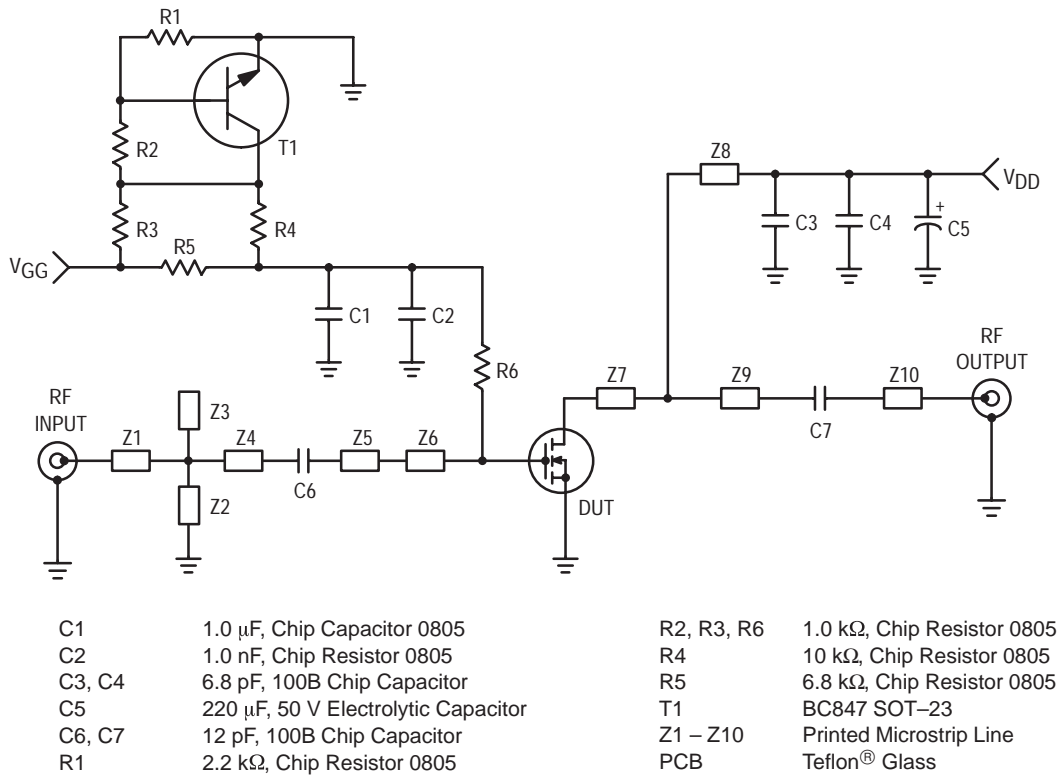


Figure 1. 1.93 – 1.99 MHz Test Fixture Schematic

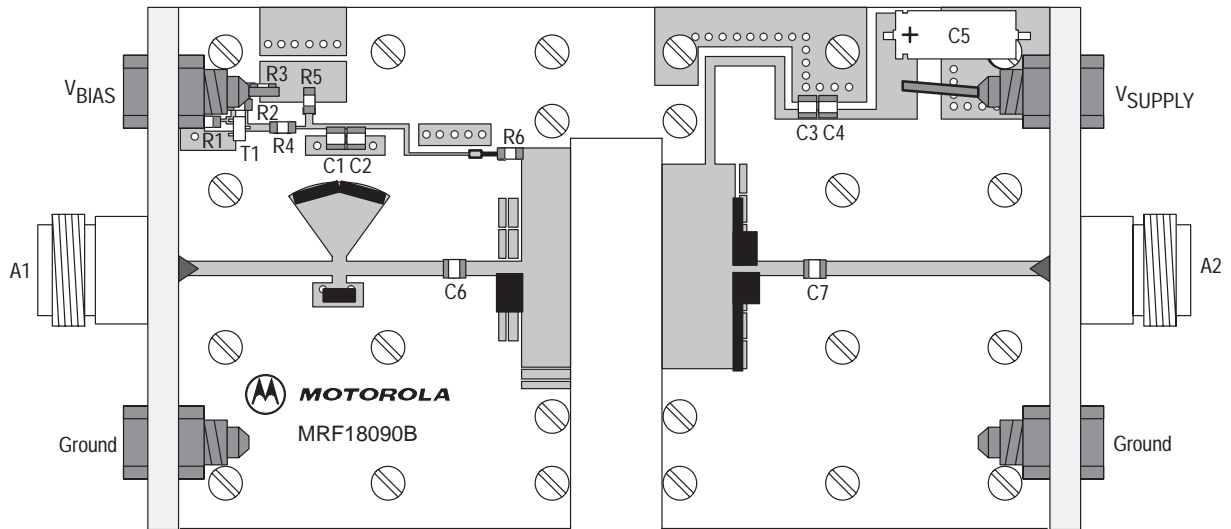
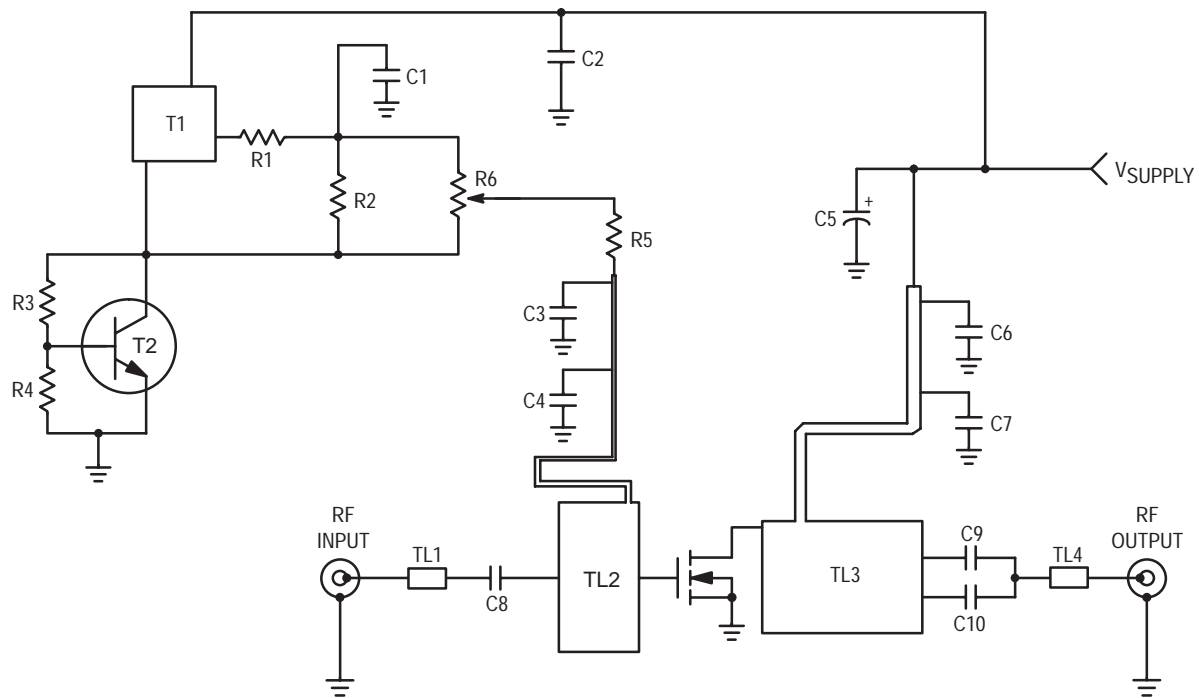


Figure 2. 1.93 – 1.99 GHz Test Fixture Component Layout



| | | | |
|-------------|--|-----------|--|
| C1, C3 | 1 μ F, Chip Capacitor 0805 | R1 | 10 Ω , Chip Resistor 0805 |
| C2 | 0.1 μ F, Chip Capacitor 0805 | R2, R3 | 1 k Ω , Chip Resistor 0805 |
| C4 | 1 nF, Chip Capacitor 0805 | R4 | 2.2 k Ω , Chip Resistor 0805 |
| C5 | 220 μ F, 50 V Electrolytic Capacitor | R5 | 10 k Ω , Chip Resistor 0603 |
| C6, C7 | 8.2 pF, 100A Chip Capacitor | R6 | 5 k Ω , SMD Potentiometer |
| C8, C9, C10 | 22 pF, 100A Chip Capacitor | T1 | LP2951 Micro-8 Voltage Regulator |
| | | T2 | BC847 SOT-23 NPN Transistor |
| | | TL1 – TL4 | Printed Transmission Lines |
| | | | Substrate = 0.5 mm Teflon [®] Glass |

Figure 3. 1.93 – 1.99 GHz Demo Board Schematic

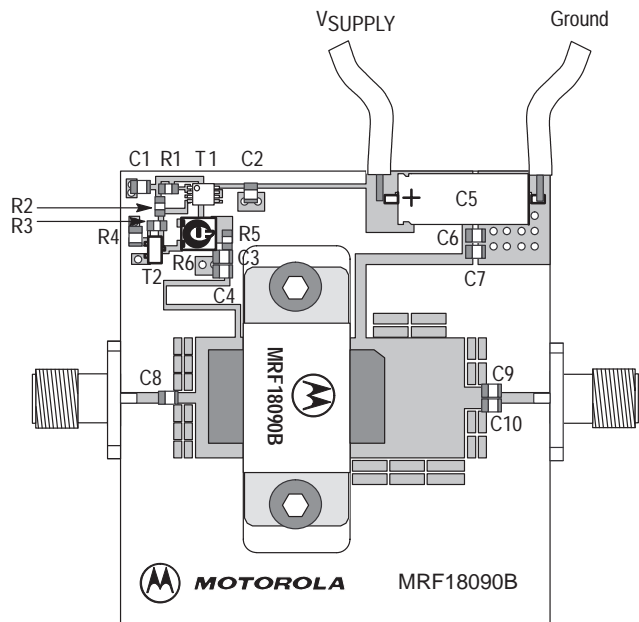


Figure 4. 1.93 – 1.99 GHz Demo Board Component Layout

TYPICAL CHARACTERISTICS

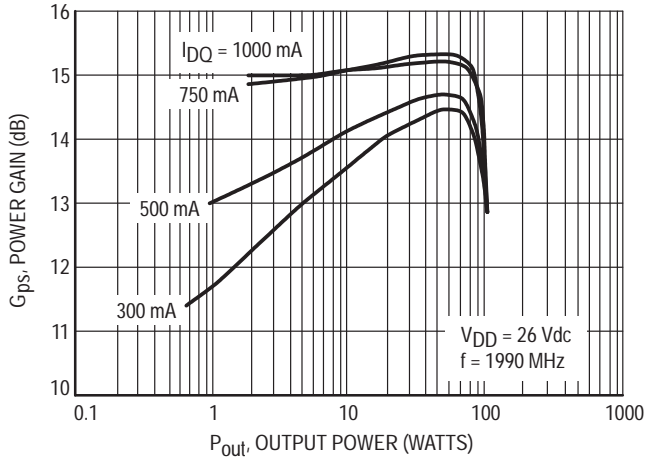


Figure 5. Power Gain versus Output Power

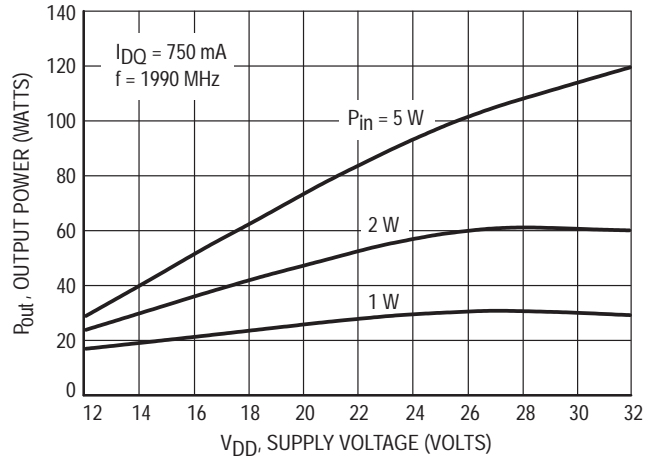


Figure 6. Output Power versus Supply Voltage

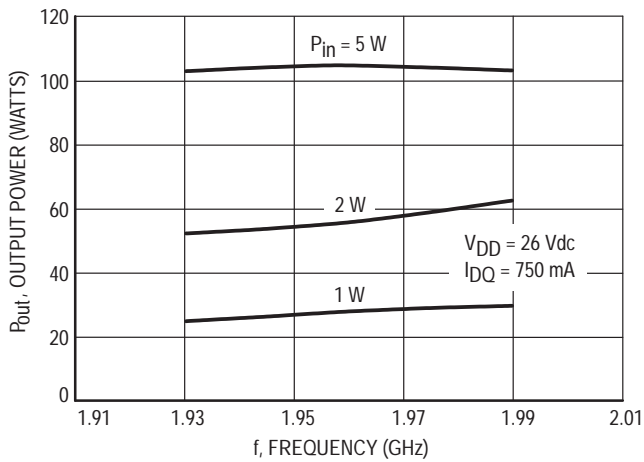


Figure 7. Output Power versus Frequency

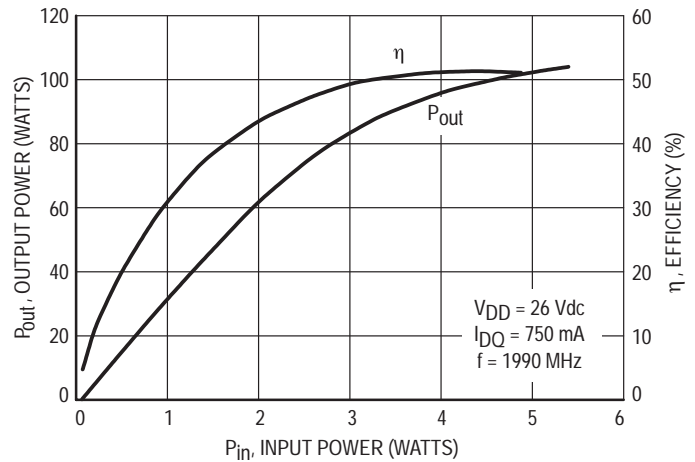


Figure 8. Output Power and Efficiency versus Input Power

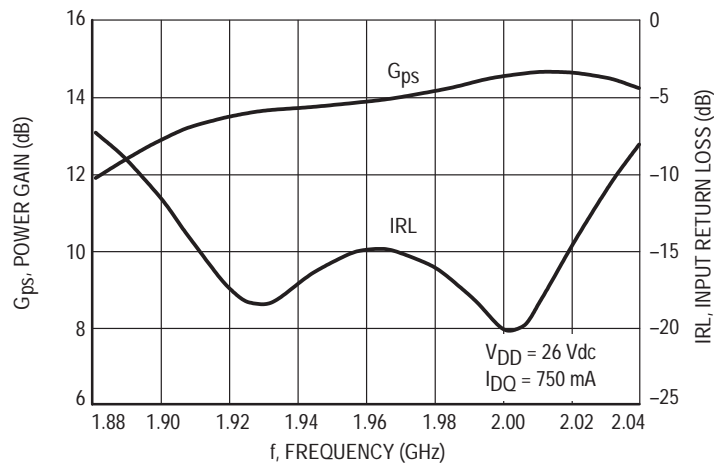


Figure 9. Wideband Gain and IRL (at Small Signal)

$V_{DD} = 26 \text{ V}$, $I_{DQ} = 750 \text{ mA}$, $P_{out} = 90 \text{ Watts (CW)}$

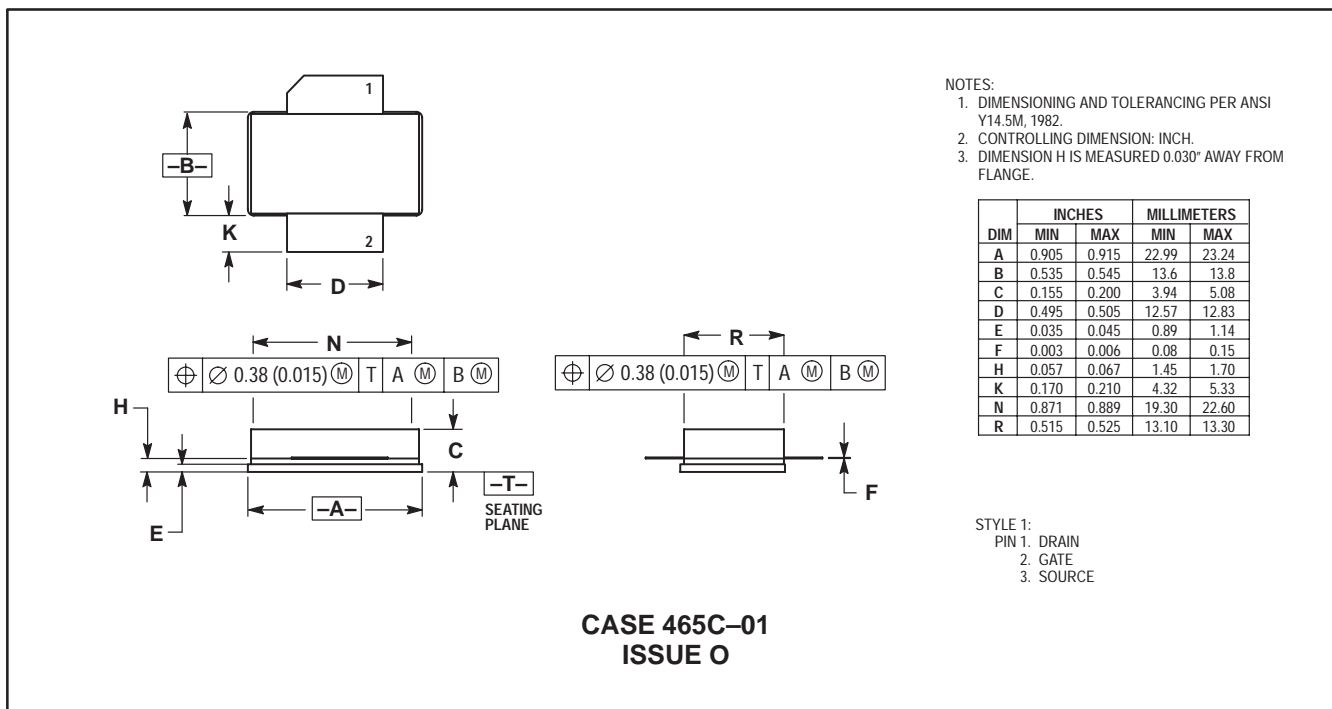
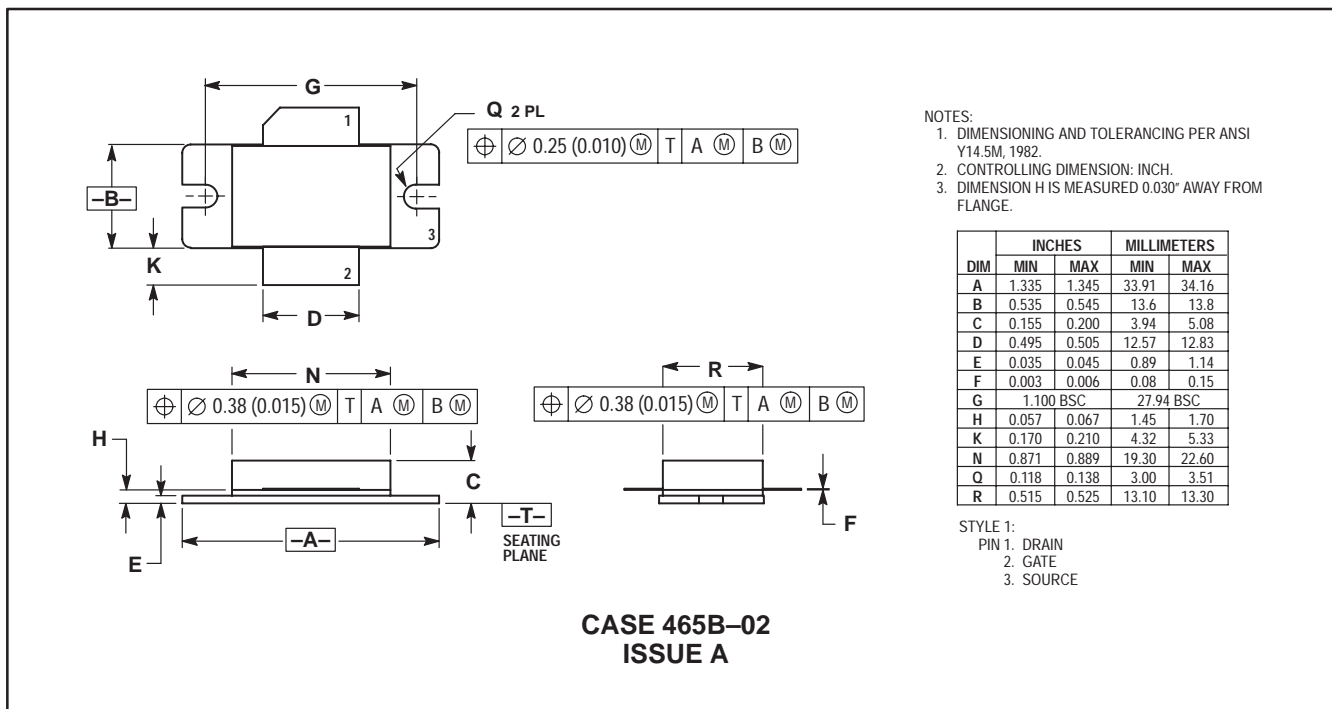
| f MHz | Z_{in} Ω | Z_{OL}^* Ω |
|------------------|--|--|
| 1805 | $1.1 + j5.85$ | $1.15 + j2.16$ |
| 1880 | $1.56 + j6.75$ | $1.13 + j2.6$ |
| 1930 | $2.05 + j8.0$ | $1.30 + j2.23$ |
| 1990 | $2.3 + j7.3$ | $0.82 + j2.90$ |


Z_{in} = Complex conjugate of source impedance.

Z_{OL}^* = Complex conjugate of the optimum load at a given voltage, P1dB, gain, efficiency, bias current and frequency.

Table 1. Large Signal Input and Output Impedance

PACKAGE DIMENSIONS



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